

2. Linear Minimum Mean Squared Error Estimation

2.1. Linear minimum mean squared error estimators

- Situation considered:**

- A random sequence $X(1), \dots, X(M)$ whose realizations can be observed.
- A random variable Y which has to be estimated.
- We seek an estimate of Y with a **linear estimator** of the form:

$$\hat{Y} = h_0 + \sum_{m=1}^M h_m X(m)$$

- A measure of the goodness of \hat{Y} is the mean squared error (MSE):

$$\mathbf{E}[(\hat{Y} - Y)^2]$$

- Covariance and variance of random variables:**

Let U and V denote two random variables with expectation

$$\mu_U \equiv \mathbf{E}[U] \text{ and } \mu_V \equiv \mathbf{E}[V].$$

- The **covariance** of U and V is defined to be:

$$\begin{aligned} \Sigma_{UV} &\equiv \mathbf{E}[(U - \mu_U)(V - \mu_V)] \\ &= \mathbf{E}[UV] - \mu_U \mu_V \end{aligned}$$

- The **variance** of U is defined to be:

$$\begin{aligned} \sigma_U^2 &\equiv \mathbf{E}[(U - \mu_U)^2] = \Sigma_{UU} \\ &= \mathbf{E}[U^2] - (\mu_U)^2 \end{aligned}$$

Let $U \equiv [U(1), \dots, U(M)]^T$ and $V \equiv [V(1), \dots, V(M)]^T$ denote two random vectors.

The covariance matrix of U and V is defined as

$$\Sigma_{UV} \equiv \begin{bmatrix} \Sigma_{U(1)V(1)} & \dots & \Sigma_{U(1)V(M)} \\ \dots & \dots & \dots \\ \Sigma_{U(M)V(1)} & \dots & \Sigma_{U(M)V(M)} \end{bmatrix}$$

A direct way to obtain Σ_{UV} :

$$\begin{aligned} \Sigma_{UV} &= \mathbf{E}[(U - \mu_U)(V - \mu_V)^T] \\ &= \mathbf{E}[UV^T] - \mu_U(\mu_V)^T \end{aligned}$$

where

$$\begin{aligned} \mu_U &\equiv \mathbf{E}[U] = [\mathbf{E}[U(1)], \dots, \mathbf{E}[U(M)]]^T \\ \mu_V &\equiv \mathbf{E}[V] \end{aligned}$$

Examples: $U = X \equiv [X(1), \dots, X(M)]^T$ and $V = Y$.

In the sequel we shall frequently make use of the following covariance matrix and vector:

$$\begin{aligned} \text{(i) } \Sigma_{XX} &= \mathbf{E}[(X - \mu_X)(X - \mu_X)^T] \\ &= \begin{bmatrix} \sigma_{X(1)}^2 & \dots & \Sigma_{X(1)X(M)} \\ \dots & \dots & \dots \\ \Sigma_{X(M)X(1)} & \dots & \sigma_{X(M)}^2 \end{bmatrix} \end{aligned}$$

$$\begin{aligned} \text{(ii) } \Sigma_{XY} &= \mathbf{E}[(X - \mu_X)(Y - \mu_Y)] \\ &= [\Sigma_{X(1)Y} \dots \Sigma_{X(M)Y}]^T \end{aligned}$$

- Linear minimum mean squared error estimator (LMMSEE)**

A LMMSEE of Y is a linear estimator, i.e. an estimator of the form

$$\hat{Y} = h_0 + \sum_{m=1}^M h_m X(m)$$

which minimizes the MSE $\mathbf{E}[(\hat{Y} - Y)^2]$.

A linear estimator is entirely determined by the $(M + 1)$ -dimensional vector

$$\mathbf{h} \equiv [h_0, \dots, h_M]^T.$$

• **Orthogonality principle:**

Orthogonality principle:

A necessary condition for $\mathbf{h} \equiv [h_0, \dots, h_M]^T$ to be the coefficient vector of the LMMSEE is that its entries fulfils the $(M + 1)$ identities:

$$\mathbf{E}[Y - \hat{Y}] = \mathbf{E}\left[Y - \left(h_0 + \sum_{m=1}^M h_m X(m)\right)\right] = 0 \quad (2.1a)$$

$$\mathbf{E}[(Y - \hat{Y})X(j)] = \mathbf{E}\left[\left\{Y - \left(h_0 + \sum_{m=1}^M h_m X(m)\right)\right\}X(j)\right] = 0, \quad (2.1b)$$

$$j = 1, \dots, M$$

Proof:

Because the coefficient vector of the LMMSEE minimizes $\mathbf{E}[(\hat{Y} - Y)^2]$, its components must satisfy the set of equations:

$$\frac{\partial}{\partial h_j} \mathbf{E}[(\hat{Y} - Y)^2] = 0 \quad j = 0, \dots, M.$$

□

Notice that the two expressions in (2.1) can be rewritten as:

$$\mathbf{E}[Y - \hat{Y}] = 0 \quad (2.2a)$$

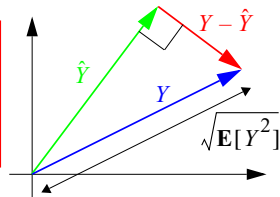
$$\mathbf{E}[(Y - \hat{Y})X(j)] = 0 \quad j = 1, \dots, M \quad (2.2b)$$

Important consequences of the orthogonality principle:

$$\mathbf{E}[(Y - \hat{Y})\hat{Y}] = 0 \quad (2.3a)$$

$$\mathbf{E}[(Y - \hat{Y})^2] = \mathbf{E}[Y^2] - \mathbf{E}[\hat{Y}^2] \quad (2.3b)$$

$$= \mathbf{E}[(Y - \hat{Y})Y]$$



□

Geometrical interpretation:

Let U and V denote two random variables with finite second moment, i.e.

$$\mathbf{E}[U^2] < \infty \text{ and } \mathbf{E}[V^2] < \infty.$$

Then, the expectation $\mathbf{E}[UV]$ can be viewed as the **scalar or inner product** of U and V .

Within this interpretation:

- U and V are uncorrelated, i.e. $\mathbf{E}[UV] = 0$ if and only if, they are **orthogonal**,
- $\sqrt{\mathbf{E}[U^2]}$ is the norm (length) of U .

Interpretation of both equations in (2.3):

- (2.3a): the estimation error $Y - \hat{Y}$ and the estimate \hat{Y} are orthogonal.
- (2.3b): results from Pythagoras' Theorem.

• **Computation of the coefficient vector of the LMMSEE:**

The coefficients of the LMMSEE satisfy the relationships:

$$\mu_Y = h_0 + \sum_{m=1}^M h_m \mu_{X(m)} = h_0 + (\mathbf{h}^-)^T \mu_X$$

$$\Sigma_{XY} = \Sigma_{XX} \mathbf{h}^-$$

where $\mathbf{h}^- \equiv [h_1, \dots, h_M]^T$ and $\mathbf{X} \equiv [X_1, \dots, X_M]^T$.

Proof:

Both identities follow by appropriately reformulating relations (2.1a) and (2.1b) and using a matrix notation for the latter one.

Thus, provided $(\Sigma_{XX})^{-1}$ exists the coefficients of the LMMSEE are given by:

$$\mathbf{h}^- = (\Sigma_{XX})^{-1} \Sigma_{XY} \quad (2.4a)$$

$$h_0 = \mu_Y - (\mathbf{h}^-)^T \mu_X = \mu_Y - \Sigma_{XY}^T (\Sigma_{XX})^{-1} \mu_X \quad (2.4b)$$

• **Example: Linear prediction of a WSS process**

Let $Y(n)$ denote a WSS process with

- zero mean, i.e. $\mathbf{E}[Y(n)] = 0$,
- autocorrelation function $\mathbf{E}[Y(n)Y(n+k)] = R_{YY}(k)$

We seek the LMMSEE for the present value of $Y(n)$ based on the M past observations $Y(n-1), \dots, Y(n-M)$ of the process. Hence,

- $Y = Y(n)$
- $X(m) = Y(n-m), m = 1, \dots, M$, i.e.

$$\mathbf{X} = [Y(n-1), \dots, Y(n-M)]^T$$

Because $\mu_Y = 0$ and $\mu_X = 0$, it follows from (2.4b) that

$$h_0 = 0$$

Computation of Σ_{XY} and Σ_{XX} :

$$-\Sigma_{XY} = [\mathbf{E}[Y(n-1)Y(n)], \dots, \mathbf{E}[Y(n-M)Y(n)]]^T$$

$$= [R_{YY}(1), \dots, R_{YY}(M)]^T$$

$$-\Sigma_{XX} =$$

$$= \begin{bmatrix} \mathbf{E}[Y(n-1)^2] & \mathbf{E}[Y(n-1)Y(n-2)] & \dots & \mathbf{E}[Y(n-1)Y(n-M)] \\ \mathbf{E}[Y(n-2)Y(n-1)] & \mathbf{E}[Y(n-2)^2] & \dots & \mathbf{E}[Y(n-2)Y(n-M)] \\ \dots & \dots & \dots & \dots \\ \mathbf{E}[Y(n-M)Y(n-1)] & \mathbf{E}[Y(n-M)Y(n-2)] & \dots & \mathbf{E}[Y(n-M)^2] \end{bmatrix}$$

$$= \begin{bmatrix} R_{YY}(0) & R_{YY}(1) & R_{YY}(2) & \dots & R_{YY}(M-1) \\ R_{YY}(1) & R_{YY}(0) & R_{YY}(1) & \dots & R_{YY}(M-2) \\ R_{YY}(2) & R_{YY}(1) & R_{YY}(0) & \dots & R_{YY}(M-3) \\ \dots & \dots & \dots & \dots & \dots \\ R_{YY}(M-1) & R_{YY}(M-2) & R_{YY}(M-3) & \dots & R_{YY}(0) \end{bmatrix}$$

□

• **Residual error using a LMMSEE:**

The MSE resulting when using a LMMSEE is

$$\mathbf{E}[(\hat{Y} - Y)^2] = \sigma_Y^2 - (\mathbf{h}^-)^T \Sigma_{XY} \quad (2.5)$$

Proof:

$$\begin{aligned} \mathbf{E}[(Y - \hat{Y})^2] &= \mathbf{E}[(Y - \hat{Y})Y] \\ &= \mathbf{E}[Y^2] - \mathbf{E}[\hat{Y}Y] \end{aligned}$$

2.2. Minimum mean squared error estimators

• **Conditional expectation:**

Let U and V denote two random variables.

The conditional expectation of V given $U = u$ is observed is defined to be

$$\mathbf{E}[V|u] \equiv \int v p(v|u) dv.$$

Notice that $\mathbf{E}[V|U]$ is a random variable. In the sequel we shall make use of the following important property of conditional expectations:

$$\mathbf{E}[\mathbf{E}[V|U]] = \mathbf{E}[V]$$

Proof:

□

• **Minimum mean squared error estimator (MMSEE):**

The MMSEE of Y based on the observation of $X(1), \dots, X(M)$ is of the form:

$$\widehat{Y}(X(1), \dots, X(M)) = \mathbf{E}[Y|X(1), \dots, X(M)]$$

Hence if $X(1) = x(1), \dots, X(M) = x(M)$ is observed, then

$$\begin{aligned} \widehat{Y}(x(1), \dots, x(M)) &= \mathbf{E}[Y|x(1), \dots, x(M)] \\ &= \int y p(y|x(1), \dots, x(M)) dy \end{aligned}$$

Proof:

Let \hat{Y} denote an arbitrary estimator. Then,

$$\begin{aligned} \mathbf{E}[(\hat{Y} - Y)^2] &= \mathbf{E}[(\hat{Y} - \widehat{Y}) - (Y - \widehat{Y})]^2 \\ &= \mathbf{E}[(\hat{Y} - \widehat{Y})^2] - \underbrace{2\mathbf{E}[(\hat{Y} - \widehat{Y})(Y - \widehat{Y})]}_{=0} + \mathbf{E}[(Y - \widehat{Y})^2] \\ &= \mathbf{E}[(\hat{Y} - \widehat{Y})^2] + \mathbf{E}[(Y - \widehat{Y})^2] \end{aligned}$$

Thus,

$$\mathbf{E}[(\hat{Y} - Y)^2] \geq \mathbf{E}[(Y - \widehat{Y})^2]$$

with equality if, and only if, $\hat{Y} = \widehat{Y}$. We still have to prove that

$$\mathbf{E}[(\hat{Y} - \widehat{Y})(Y - \widehat{Y})] = 0.$$

□

Example: Multivariate Gaussian variables:

$[Y, X(1), \dots, X(M)]^T \sim N(\mu, \Sigma)$ with

$$\begin{aligned} -\mu &\equiv [\mu_Y, \mu_{X(1)}, \dots, \mu_{X(M)}]^T \\ -\Sigma &\equiv \begin{bmatrix} \sigma_Y^2 & (\Sigma_{XY})^T \\ \Sigma_{XY} & \Sigma_{XX} \end{bmatrix} \end{aligned}$$

From Equation (6.22) in [Shanmugan] it follows that

$$\widehat{Y} = \mathbf{E}[Y|X] = \mu_Y + (\Sigma_{XY})^T (\Sigma_{XX})^{-1} (X - \mu_X)$$

Bivariate case: $M = 1, X(1) = X$

$$-\Sigma_{XX} = \sigma_X^2,$$

$$-\Sigma_{XY} = \rho \sigma_X \sigma_Y, \text{ where } \rho \equiv \frac{\Sigma_{XY}}{\sigma_X \sigma_Y} \text{ is the correlation coefficient of } Y \text{ and } X.$$

In this case,

$$\begin{aligned} \widehat{Y} = \mathbf{E}[Y|X] &= \mu_Y + \frac{\rho \sigma_Y}{\sigma_X} (X - \mu_X) \\ &= \underbrace{\left(\mu_Y - \frac{\rho \sigma_Y}{\sigma_X} \mu_X \right)}_{h_0} + \underbrace{\left(\frac{\rho \sigma_Y}{\sigma_X} \right)}_{h_1} X \end{aligned}$$

We can observe that \widehat{Y} is linear, i.e. is the LMMSEE $\hat{Y} = \widehat{Y}$ in the bivariate case. This is also true in the general multivariate Gaussian case. In fact,

$$\hat{Y} = \widehat{Y} \text{ if, and only if, } [Y, X(1), \dots, X(M)]^T \text{ is a Gaussian random vector.}$$

2.3. Time-discrete Wiener filters

• **Problem:**

Estimation of a WSS random sequence $Y(n)$ based on the observation of another sequence $X(n)$. Without loss of generality we assume that $\mathbf{E}[Y(n)] = \mathbf{E}[X(n)] = 0$.

The goodness of the estimator $\hat{Y}(n)$ is described by the MSE

$$\mathbf{E}[(\hat{Y}(n) - Y(n))^2].$$

We distinguish between two cases:

- **Prediction:**

$\hat{Y}(n)$ depends on one or several past observations of $X(n)$ only, i.e.

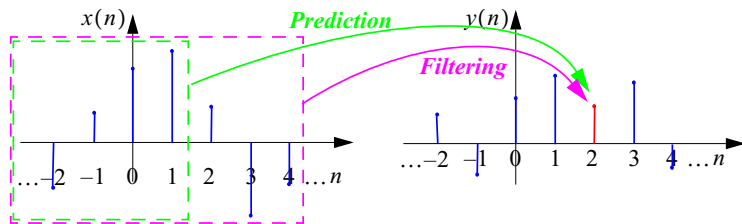
$$\hat{Y}(n) = \hat{Y}(X(n_1), X(n_2), \dots) \text{ with } n_1, n_2, \dots < n$$

- **Filtering:**

$\hat{Y}(n)$ depends on the present observation and/or one or many future observation(s) of $X(n)$, i.e.

$$\hat{Y}(n) = \hat{Y}(X(n_1), X(n_2), \dots) \text{ where at least one } n_i \geq n$$

If all $n_i \leq n$, the filter is **causal** otherwise it is **noncausal**.



Typical application: WSS signal embedded in additive white noise

$$X(n) = Y(n) + W(n) .$$

where,

- $W(n)$ is a white noise sequence,
- $Y(n)$ is a WSS process
- $Y(n)$ and $W(n)$ are uncorrelated.

However, the theoretical treatment is more general as shown below.

2.3.1. Noncausal Wiener filters

• **Linear Minimum Mean Squared Error Filter**

We seek a linear filter

$$\hat{Y}(n) = \sum_{m=-\infty}^{\infty} h(m)X(n-m) = h(n)*X(n)$$

which minimizes the MSE $\mathbf{E}[(\hat{Y}(n) - Y(n))^2]$.

Such a filter exists. It is called a **Wiener filter** in honour of his inventor.

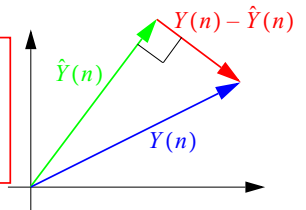
• **Orthogonality principle (time domain):**

The coefficients of a Wiener filter satisfy the conditions:

$$\begin{aligned} \mathbf{E}[(Y(n) - \hat{Y}(n))X(n-k)] &= \\ &= \mathbf{E}\left[\left(Y(n) - \sum_{m=-\infty}^{\infty} h(m)X(n-m)\right)X(n-k)\right] = 0, \quad k = \dots, -1, 0, 1, \dots \end{aligned}$$

It follows from these identities (see also (2.3)) that

$$\begin{aligned} \mathbf{E}[(Y(n) - \hat{Y}(n))\hat{Y}(n)] &= 0 \\ \mathbf{E}[(Y(n) - \hat{Y}(n))^2] &= \mathbf{E}[Y(n)^2] - \mathbf{E}[\hat{Y}(n)^2] \\ &= \mathbf{E}[(Y(n) - \hat{Y}(n))Y(n)] \end{aligned}$$



With the definitions

$$R_{XX}(k) \equiv \mathbf{E}[X(n)X(n+k)], \quad R_{XY}(k) \equiv \mathbf{E}[X(n)Y(n+k)]$$

we can recast the orthogonality conditions as follows:

$$\begin{aligned} R_{XY}(k) &= \sum_{m=-\infty}^{\infty} h(m)R_{XX}(k-m) \quad k = \dots, -1, 0, 1, \dots \\ R_{XY}(k) &= h(k)*R_{XX}(k) \quad \text{Wiener-Hopf equation} \end{aligned}$$

- **Orthogonality principle (frequency domain):**

$$S_{XY}(f) = H(f)S_{XX}(f)$$

where

$$S_{XY}(f) \equiv F\{R_{XY}(k)\}$$

- **Transfer function of the Wiener filter:**

$$H(f) = \frac{S_{XY}(f)}{S_{XX}(f)}$$

- **MSE of the Wiener filter (time-domain formulation):**

$$\mathbf{E}[(Y(n) - \hat{Y}(n))^2] = \sigma_Y^2 - \sum_{m=-\infty}^{\infty} h(m)R_{XY}(m)$$

Proof:

$$\mathbf{E}[(Y(n) - \hat{Y}(n))^2] = \mathbf{E}[Y(n)^2] - \mathbf{E}[\hat{Y}(n)Y(n)]$$

□

- **MSE of the Wiener filter (frequency-domain formulation):**

We can rewrite the above identity as:

$$\mathbf{E}[(Y(n) - \hat{Y}(n))^2] = R_{YY}(0) - \sum_{m=-\infty}^{\infty} h(m)R_{YX}(-m)$$

$\mathbf{E}[(Y(n) - \hat{Y}(n))^2]$ is the value $p(0)$ of the sequence

$$\begin{aligned} p(k) &= R_{YY}(k) - \sum_{m=-\infty}^{\infty} h(m)R_{YX}(k-m) \\ &= R_{YY}(k) - h(k)*R_{YX}(k) \end{aligned}$$

$$P(f) = S_{YY}(f) - H(f)S_{YX}(f) = S_{YY}(f) - \frac{|S_{XY}(f)|^2}{S_{XX}(f)}$$

Hence,

$$\mathbf{E}[(Y(n) - \hat{Y}(n))^2] = p(0) = \int_{-1/2}^{1/2} P(f)df$$

$$\mathbf{E}[(Y(n) - \hat{Y}(n))^2] = \int_{-1/2}^{1/2} \left[S_{YY}(f) - \frac{|S_{XY}(f)|^2}{S_{XX}(f)} \right] df$$

2.3.2. Causal Wiener filters

A. $X(n)$ is a white noise.

We first assume that $X(n)$ is a white noise with unit variance, i.e.

$$\mathbf{E}[X(n)X(n+k)] = \delta(k).$$

- **Derivation of the causal Wiener filter from the noncausal Wiener filter:**

Let us consider the noncausal Wiener filter

$$\hat{Y}(n) = \sum_{m=-\infty}^{\infty} h(m)X(n-m)$$

whose transfer function is given by

$$H(f) = \frac{S_{XY}(f)}{S_{XX}(f)} = S_{XY}(f).$$

Then, the causal Wiener filter $\hat{Y}_c(n)$ results by cancelling the noncausal part of the non-causal Wiener filter:

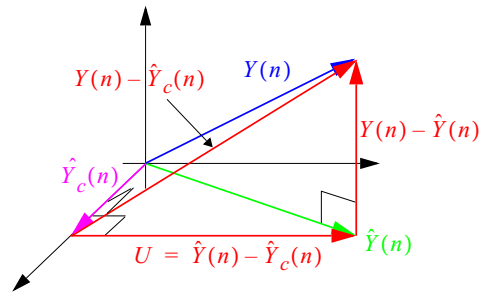
$$\hat{Y}_c(n) = \sum_{m=0}^{\infty} h(m)X(n-m)$$

Sketch of the proof:

$\hat{Y}(n)$ can be written as

$$\hat{Y}(n) = \underbrace{\sum_{m=-\infty}^{-1} h(m)X(n-m)}_{\equiv U} + \underbrace{\sum_{m=0}^{\infty} h(m)X(n-m)}_{V \equiv \hat{Y}_c(n)}$$

Because $X(n)$ is a white noise, the causal part $V = \hat{Y}_c(n)$ and the noncausal part $U = \hat{Y}(n) - \hat{Y}_c(n)$ of $\hat{Y}(n)$ are orthogonal. It follows from this property



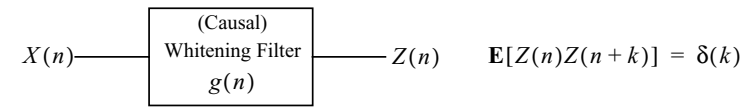
that $\hat{Y}_c(n)$ and $Y(n)$ are orthogonal, i.e. that $\hat{Y}_c(n)$ minimizes the MSE within the class of linear causal estimators. □

B. $X(n)$ is an arbitrary WSS process whose spectrum satisfies the Paley-Wiener condition.

Usually, the above truncation procedure to obtain $\hat{Y}_c(n)$ does not apply because U and V are correlated and therefore not orthogonal in the general case.

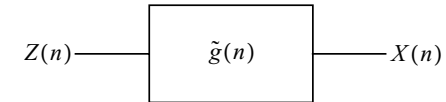
• **Causal whitening filter:**

However, we can show (see the Spectral Decomposition Theorem below) that provided $S_{XX}(f)$ satisfies the Paley-Wiener condition (see below) then $X(n)$ can be **converted into an equivalent white noise sequence $Z(n)$ with unit variance** by filtering it with an appropriate causal filter $g(n)$,



This operation is called **whitening** and the filter $g(n)$ is called a **whitening filter**.

equivalent \equiv there exists another causal filter $\tilde{g}(n)$ so that $X(n) = \tilde{g}(n)*Z(n)$:



Notice that if

$$G(f) \equiv F\{g(n)\}$$

$$\tilde{G}(f) \equiv F\{\tilde{g}(n)\}$$

then

$$|G(f)|^2 = S_{XX}(f)^{-1}$$

$$|\tilde{G}(f)|^2 = S_{XX}(f)$$

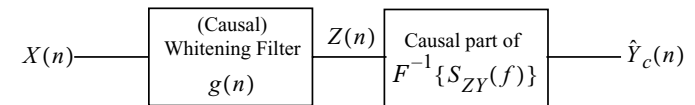
We shall see that a whitening filter exists such that

$$\tilde{G}(f) = G(f)^{-1}$$

(2.6)

• **Causal Wiener filter**

Making use of the result in Part A, the block diagram of the causal Wiener filter is



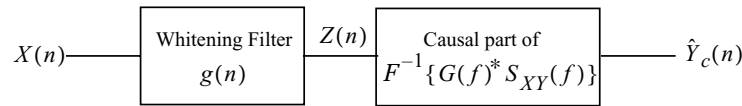
$S_{ZY}(f)$ is obtained from $S_{XY}(f)$ according to

$$S_{ZY}(f) = G(f)^* S_{XY}(f)$$

Proof:

□

Hence, the block diagram of the causal Wiener filter is:



• **Spectral Decomposition Theorem:**

Let $S_{XX}(f)$ satisfies the so-called **Paley-Wiener condition**:

$$\int_{-1/2}^{1/2} \log(S_{XX}(f))df > -\infty$$

Then $S_{XX}(f)$ can be written as

$$S_{XX}(f) = G(f)^+ G(f)^-$$

with $G(f)^+$ and $G(f)^-$ satisfying

$$|G(f)^+|^2 = |G(f)^-|^2 = S_{XX}(f)$$

Moreover, the sequences

$$\begin{aligned} g(n)^+ &\equiv F^{-1}\{G(f)^+\} \\ g(n)^- &\equiv F^{-1}\{G(f)^-\} \\ g^{-1}(n)^+ &\equiv F^{-1}\{1/G(f)^+\} \\ g^{-1}(n)^- &\equiv F^{-1}\{1/G(f)^-\} \end{aligned}$$

satisfy

$$\begin{aligned} g(n)^+ = g^{-1}(n)^+ = 0 & \quad n < 0 & \text{Causal sequences} \\ g(n)^- = g^{-1}(n)^- = 0 & \quad n > 0 & \text{Anticausal sequences} \end{aligned}$$

• **Whitening filter (cont'd):**

The sought whitening filter used to obtained $Z(n)$ is

$$g(n) = g^{-1}(n)^+$$

and

$$\tilde{g}(n) = g(n)^+$$

It can be easily verified that both sequences satisfy the identities in (2.6).

2.3.3. Finite Wiener filters

• **Finite linear filter:**

$$\hat{Y}(n) = \sum_{m=-M_1}^{M_2} h(m)X(n-m)$$

• **Wiener-Hopf equation:**

By applying the orthogonality principle we obtain the Wiener-Hopf system of equations:

$$\Sigma_{XY} = \Sigma_{XX}h$$

where

$$h \equiv [h(-M_1), \dots, h(M_2)]^T$$

and

$$\Sigma_{XY} \equiv [R_{XY}(-M_1), \dots, R_{XY}(M_2)]^T$$

$$\Sigma_{XX} \equiv$$

$$\equiv \begin{bmatrix} R_{XX}(0) & R_{XX}(1) & R_{XX}(2) & \dots & R_{XX}(M_1 + M_2) \\ R_{XX}(1) & R_{XX}(0) & R_{XX}(1) & \dots & R_{XX}(M_1 + M_2 - 1) \\ R_{XX}(2) & R_{XX}(1) & R_{XX}(0) & \dots & R_{XX}(M_1 + M_2 - 2) \\ \dots & \dots & \dots & \dots & \dots \\ R_{XX}(M_1 + M_2) & R_{XX}(M_1 + M_2 - 1) & R_{XX}(M_1 + M_2 - 2) & \dots & R_{XX}(0) \end{bmatrix}$$

Coefficient vector of the finite Wiener filter:

$$h = (\Sigma_{XX})^{-1} \Sigma_{XY}$$

provided Σ_{XX} is invertible.